

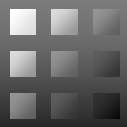
# Issues for Multi-Band Multi-Access Radio Circuits in 5G Mobile Communication

Yasushi Yamao

*AWCC*

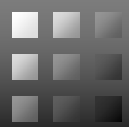
The University of Electro-Communications





# Outline

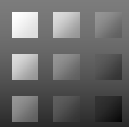
- Background
- Requirements for 5G Hardware
- Issues for Tx and Rx
- Advanced DPD techniques
- Post compensation technique for Rx
- Reconfigurable BPF
- Conclusion



# From 4G to 5G

- **5G Mobile Communication** will be changed to accept the **diversity** of ;
  - ✓ Different access protocols with wide range of spectrum from 700 MHz to more than 6 GHz (~millimeter wave)
  - ✓ Heterogeneous deployment with different cell sizes,
  - ✓ Carrier aggregation (CA) and dual access from UEs,
  - ✓ Cooperated multiple transmission (CoMP), massive MIMO and distributed antenna systems (DSA).

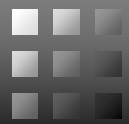
In order to achieve more efficient and flexible use of radio resources, separation of C-plane and U-plane has been studied.



# Requirements for 5G Hardware

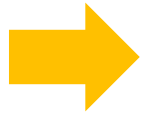
- High bandwidth of 10 Gbps or more
- Utilizing widely-spread frequency bands from current UHF to low SHF, high SHF and millimeter waves
- Low-power/cost small base station and Dual Access to both Macro- and small-cells with different frequency bands
- High Accuracy RF signals are required to increase spectral efficiency

5G hardware is required to be more **flexible**, **accurate**, **linear** and **low-cost**.



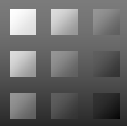
# Issues for Transmitter

- Keeping linearity and power efficiency **in a wider unit RF bandwidth of 100MHz or more.**
- Keeping linearity and power efficiency **under concurrent multiband** operation



**Nonlinear Compensation Techniques** should be developed that can work in a wider bandwidth and multi-band environments.

We proposes advanced **Digital Predistortion (DPD)** techniques called **SENF** and **SFFB**



# Issues for Receiver

- Receiver front-end faces a variety of incoming signals with power of wide dynamic range. Desired signals are not always stronger than others.
- Under **concurrent multiband** operation, near-far problem increases chance of inter-band modulation called "cross-modulation distortion".

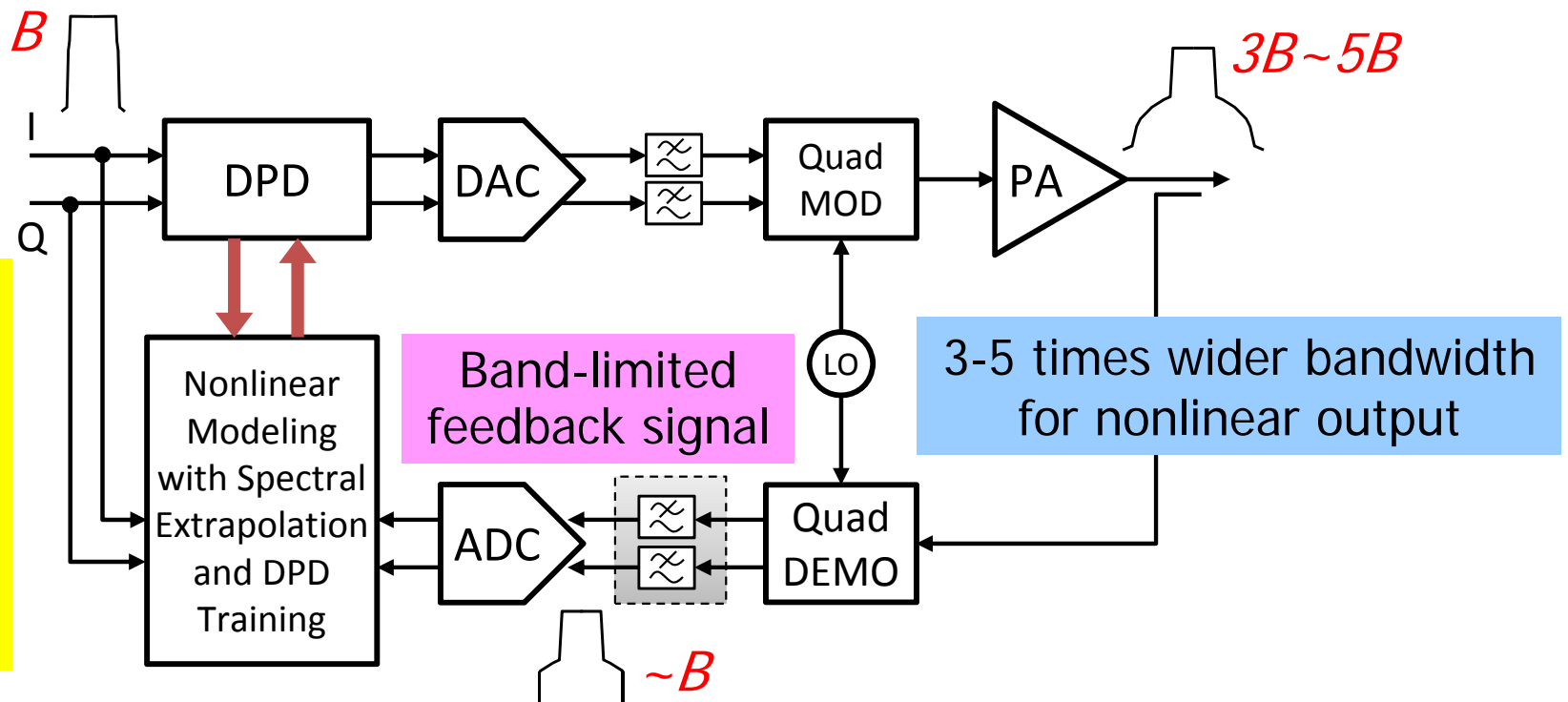


**Receiver Nonlinearity Compensation technique** and **reconfigurable RF BPF** as pre-selector mitigate the issues.

We propose a **Post Compensation technique** and **Reconfigurable BPF** for concurrent dual-band receiver.

# 1. Wideband DPD Design Method

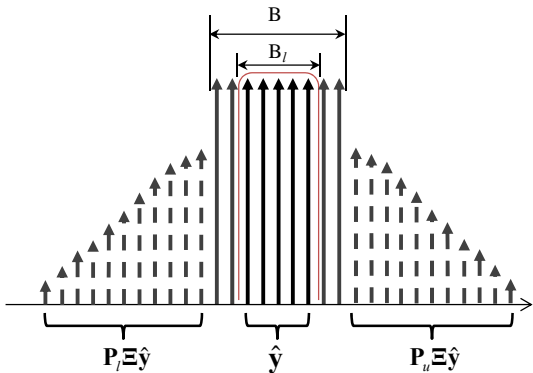
- Existing DPDs have been designed to feedback full bandwidth of nonlinear output signal, requiring 3 to 5 times wideband ADC.



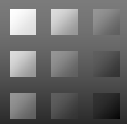
Iterated Nonlinear Modeling with Spectral Extrapolation of Narrowband Feedback signal

Band-limited feedback signal

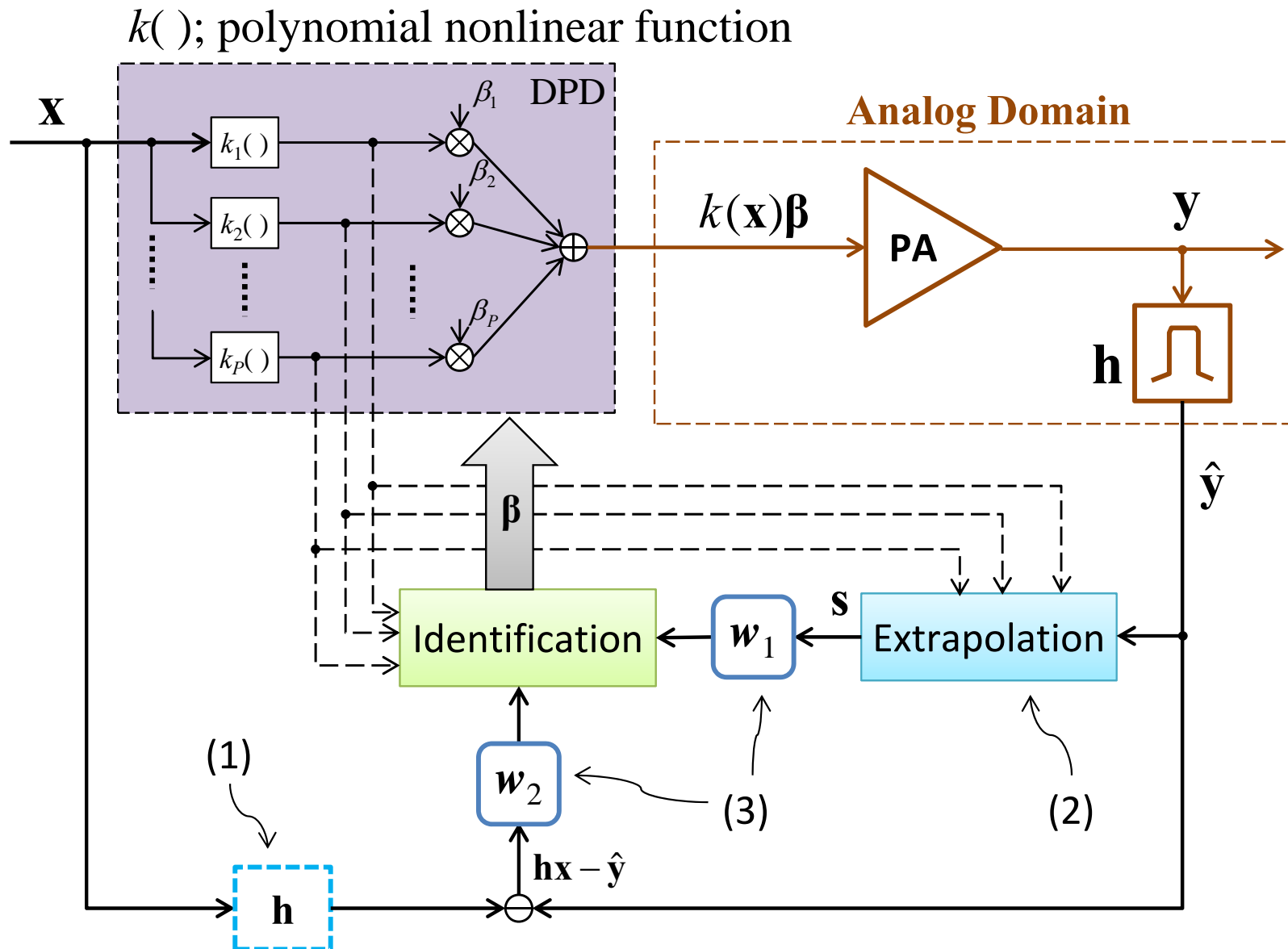
3-5 times wider bandwidth for nonlinear output



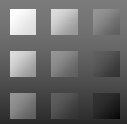
With SENF (Spectral Extrapolation of Narrowband Feedback) technique, feedback bandwidth can be same as the signal bandwidth (or even less) [1].



# SENF DPD Equivalent Baseband Diagram



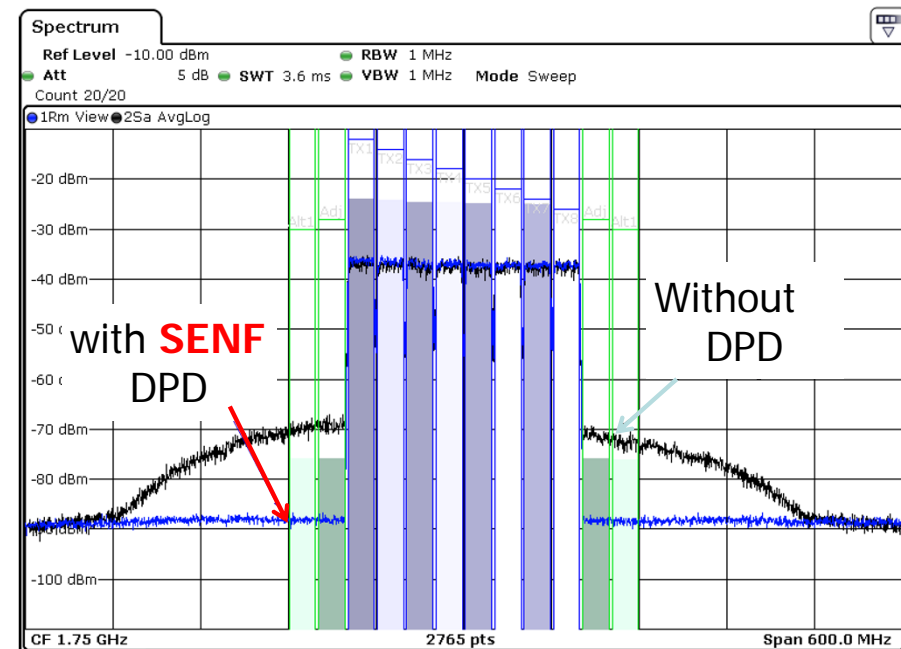
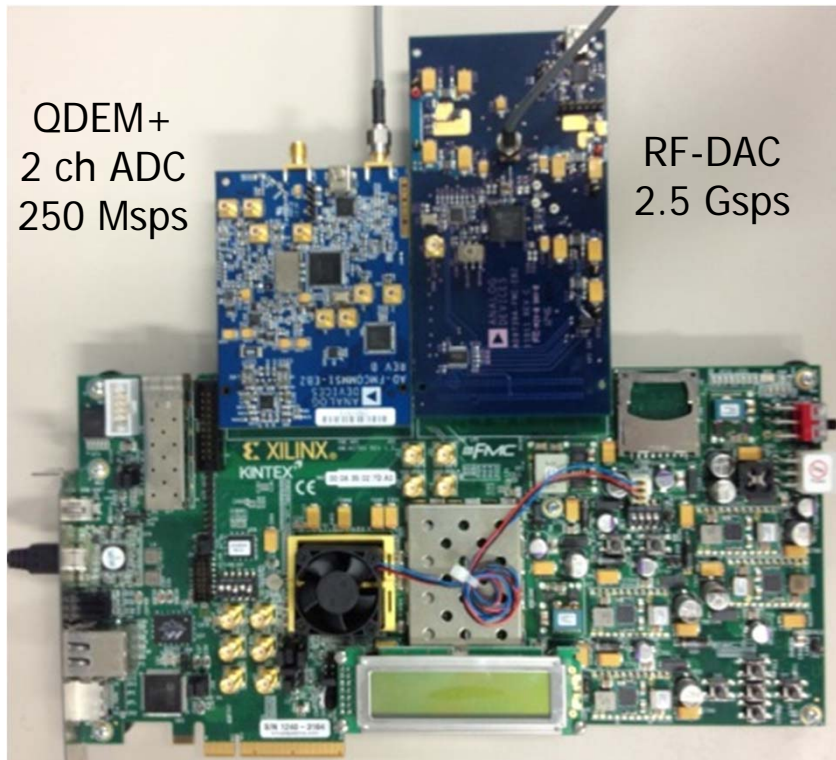




# SENF DPD by FPGA

- More than 100 MHz Linearization is possible with current FPGAs with 250 Msps ADC by **SENF method**.

RF: 1.75 GHz  
In            Out



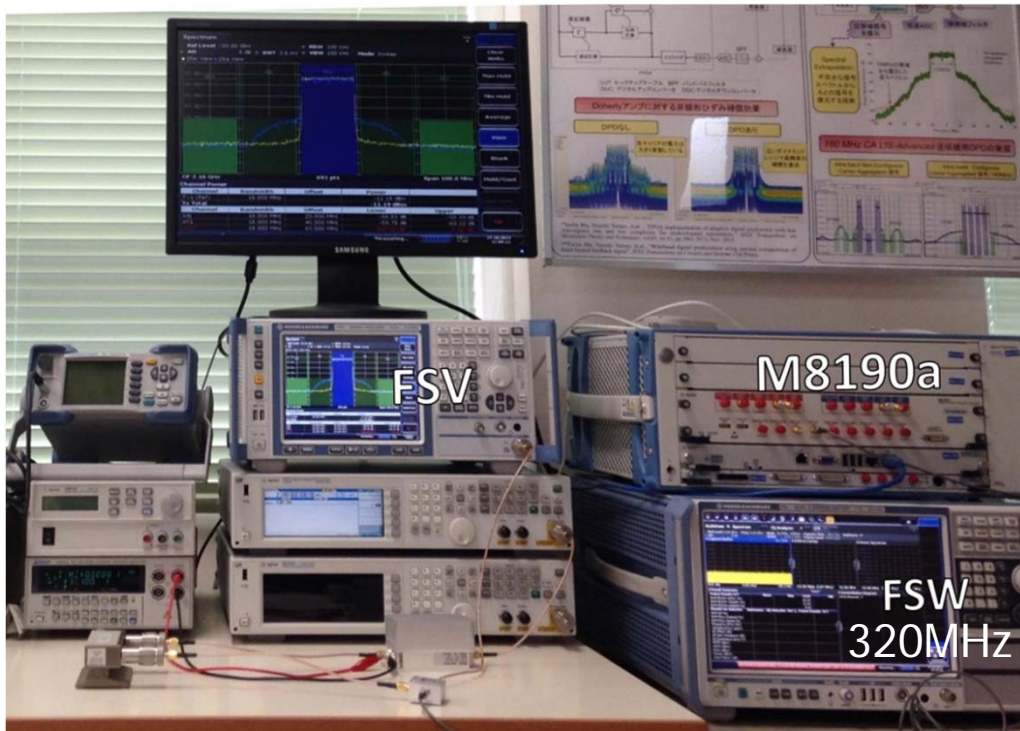
8 x 20MHz LTE CA signal (160MHz)

160MHz bandwidth DPD  
by Xilinx Kintex7 FPGA

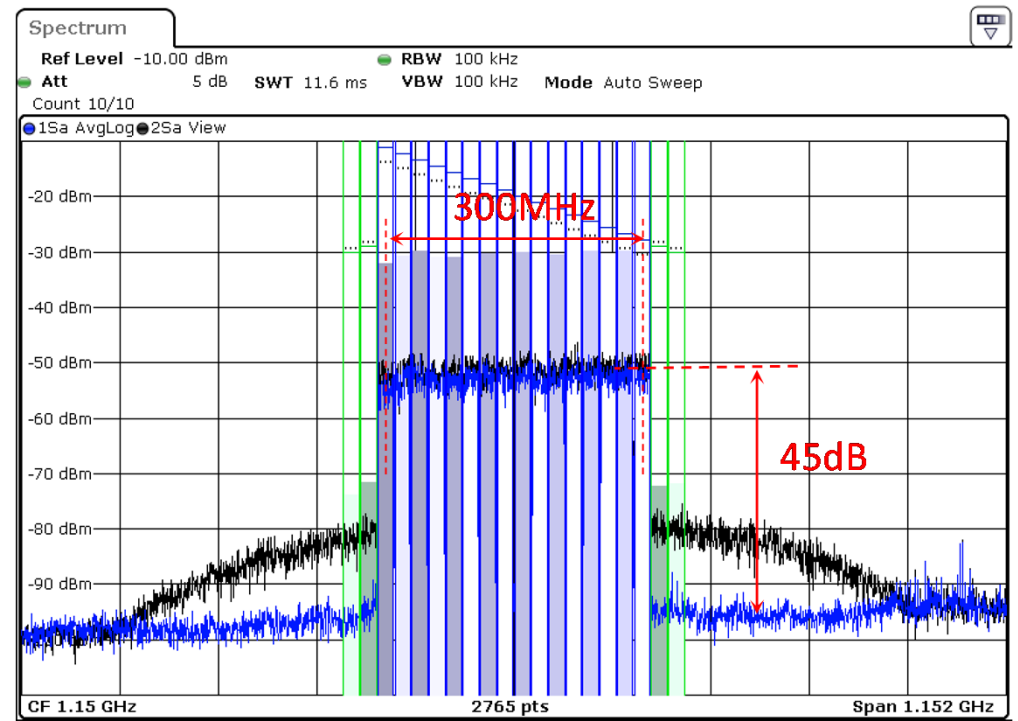


# SENF DPD by Experiment (1)

- More than 300 MHz Linearization is confirmed in Experiment [2].



320MHz feedback bandwidth DPD  
by measurement set up



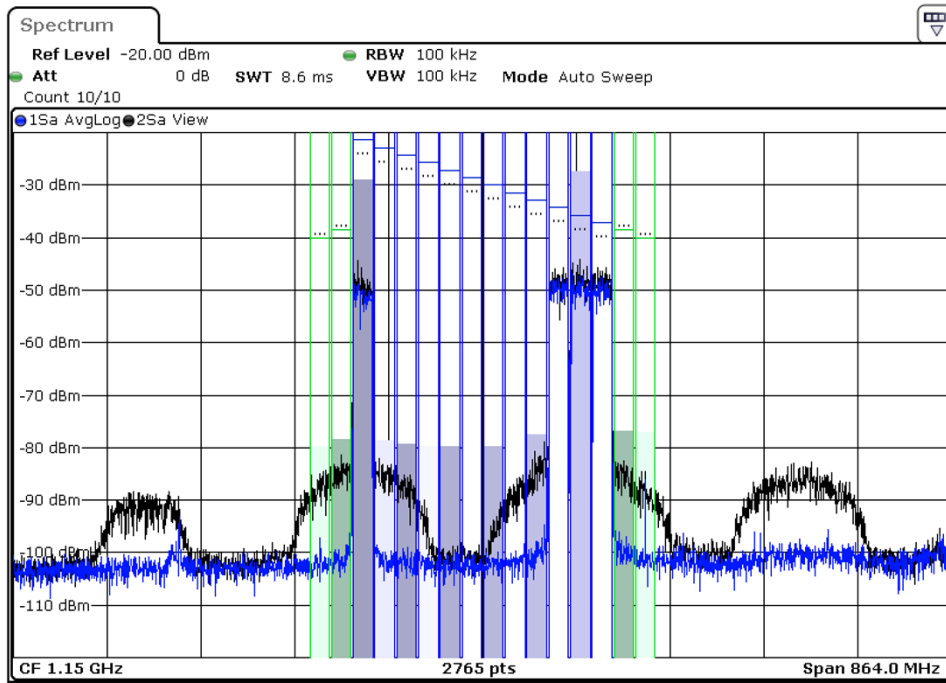
16 x 20MHz LTE CA signal (320MHz)

Linearization of **signal with 500 MHz and beyond bandwidth**  
will be achieved soon by DPD.

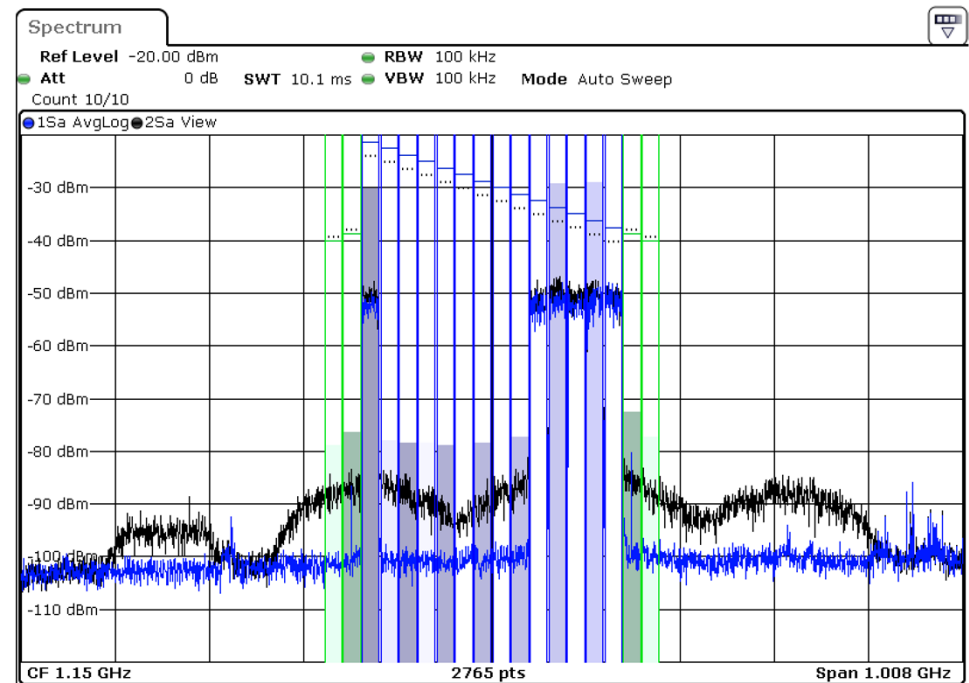


# SENF DPD by Experiment (2)

- Proposed DPD can compensate non-continuous CA signal.



4 x 20MHz LTE CA signal (240MHz)

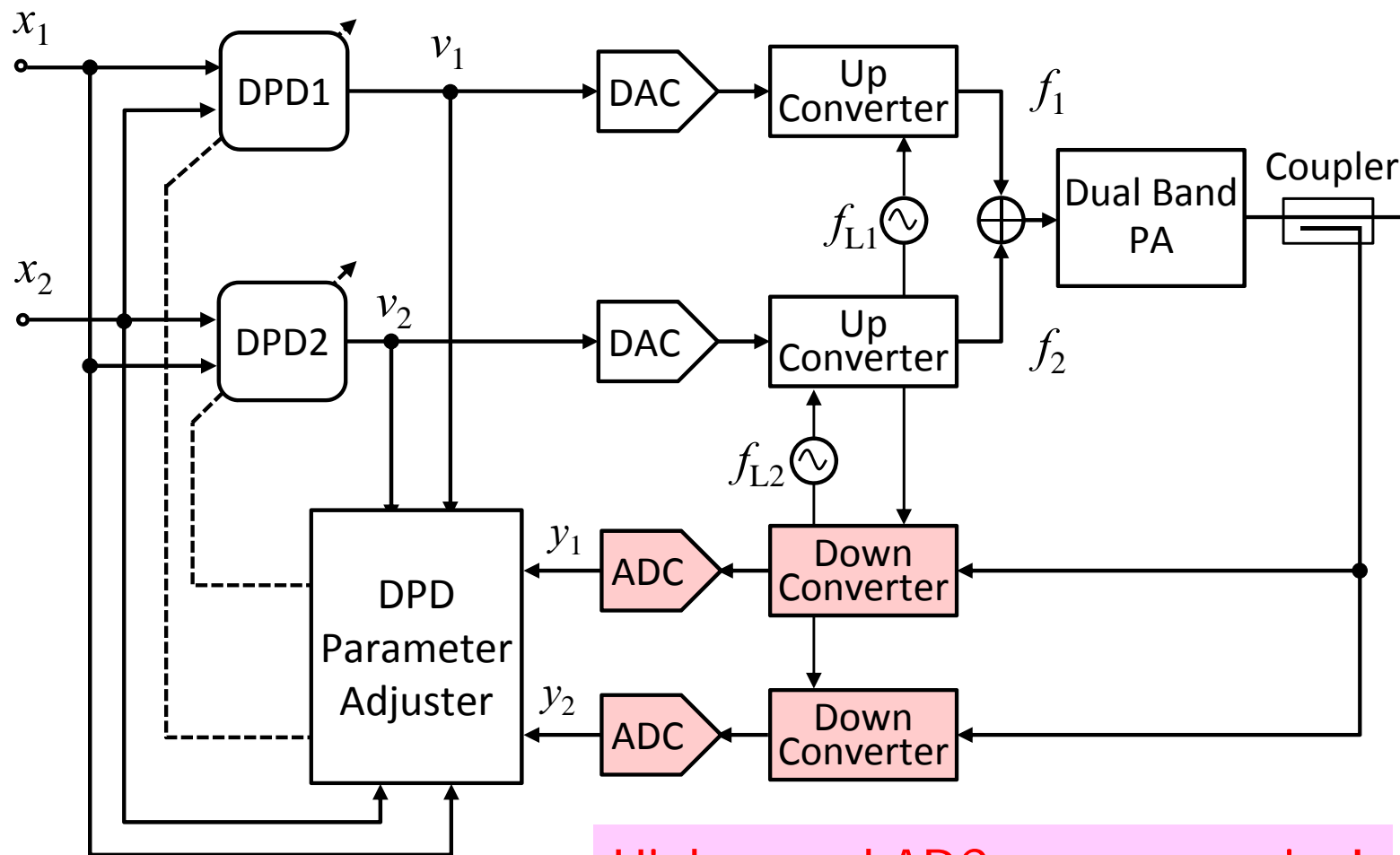


6 x 20MHz LTE CA signal (280MHz)

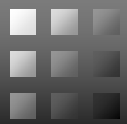
Linearization of 900 MHz and beyond RF bandwidth is achieved by SENF DPD.

## 2. Concurrent Dual-Band DPD Design

- Existing Dual-Band DPDs have two feedback path with **two sets of Down Converter and ADC**.

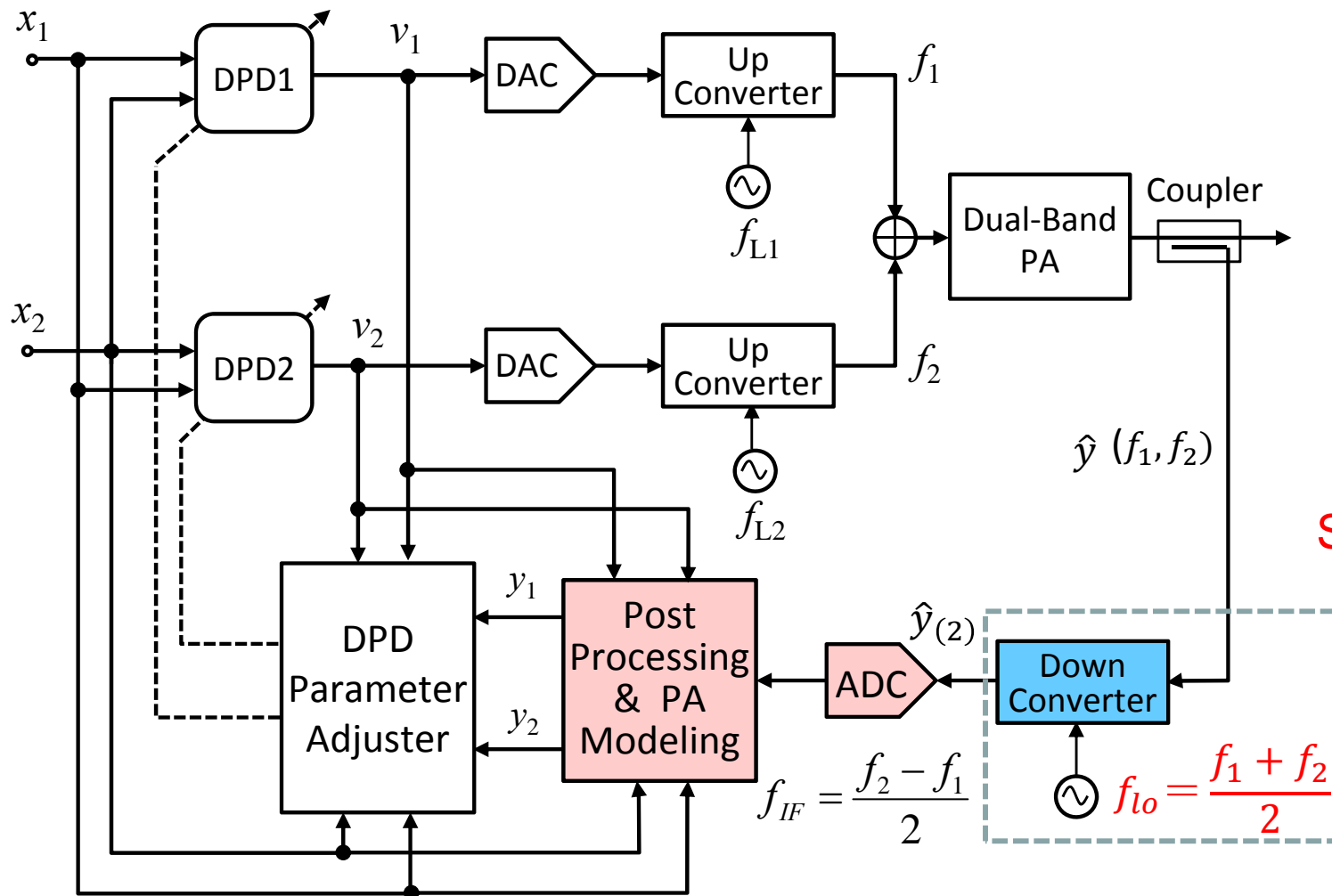


High-speed ADCs are expensive!



# SFFB Dual-Band DPD

- Spectrum Folding Feedback (SFFB) DPD multiplexes two RF spectra into common IF [3].



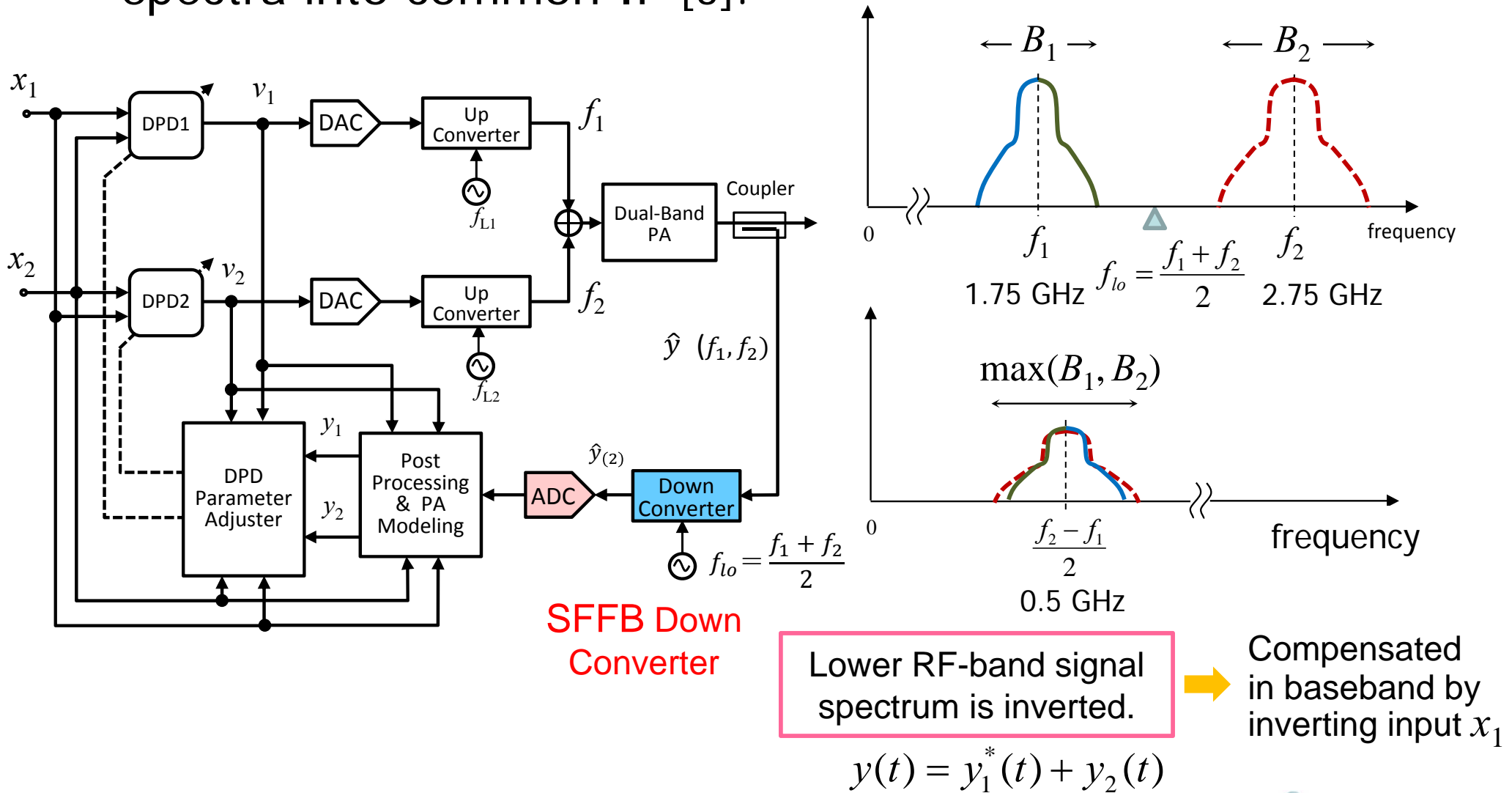
Spectrum Folding  
Down Converter

Two RF signals are  
down-converted  
into a common IF  
as a multiplexed  
signal.

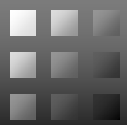


# SFFB Dual-Band DPD

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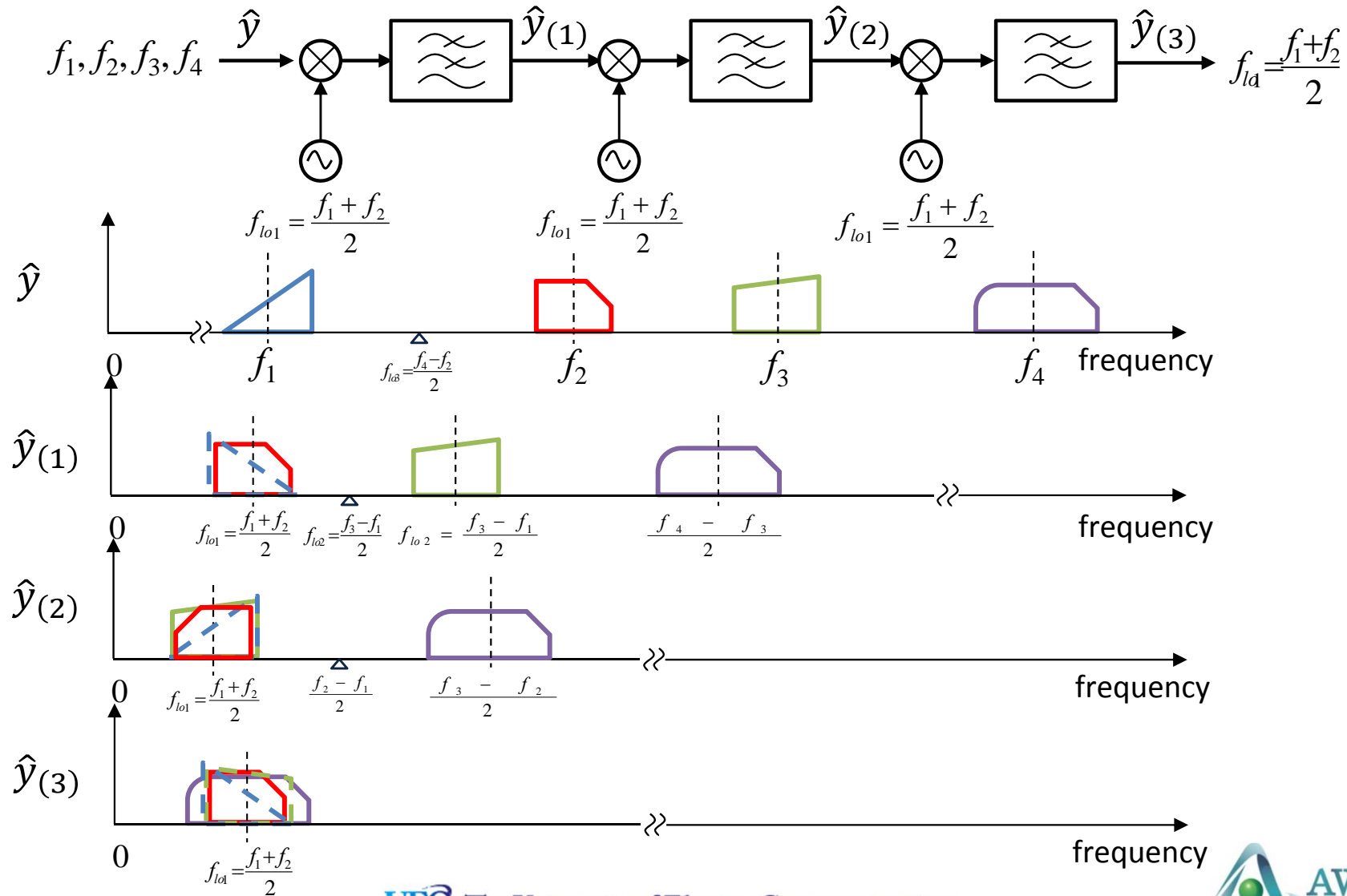


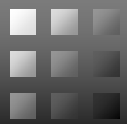




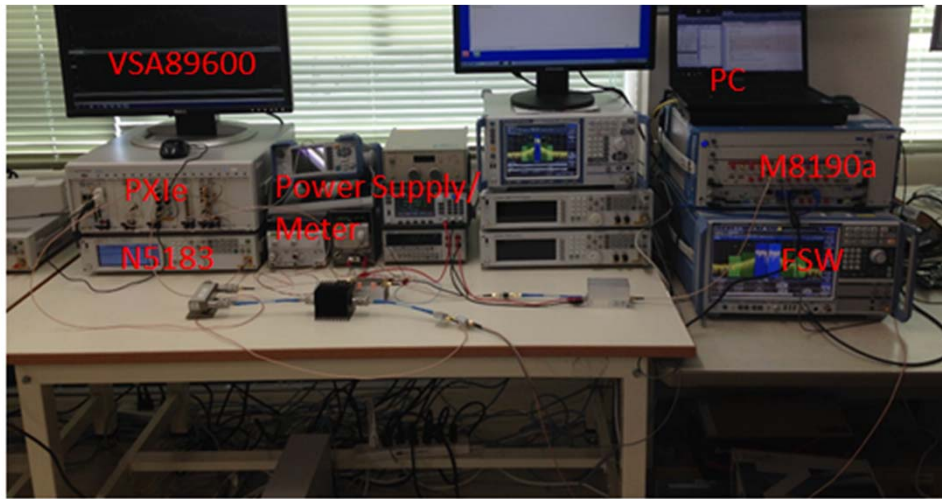
# SFFB Multi-Band Extension

With **SFFB (Spectral Folding Feedback)** technique, multiband signals can be folded into one IF bandwidth.

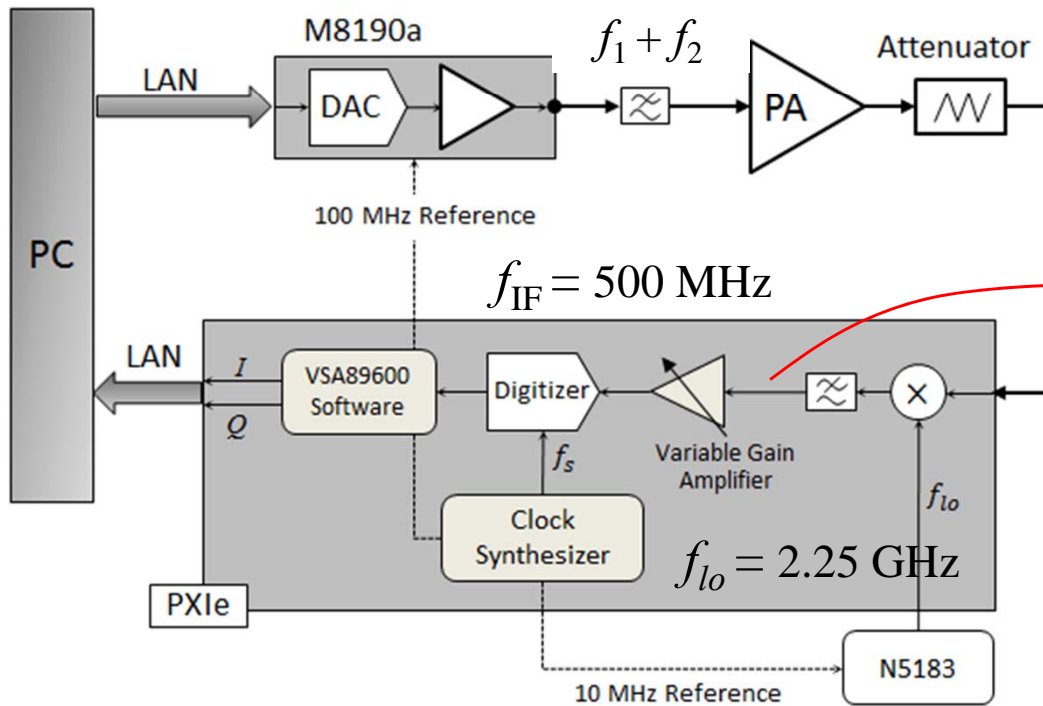
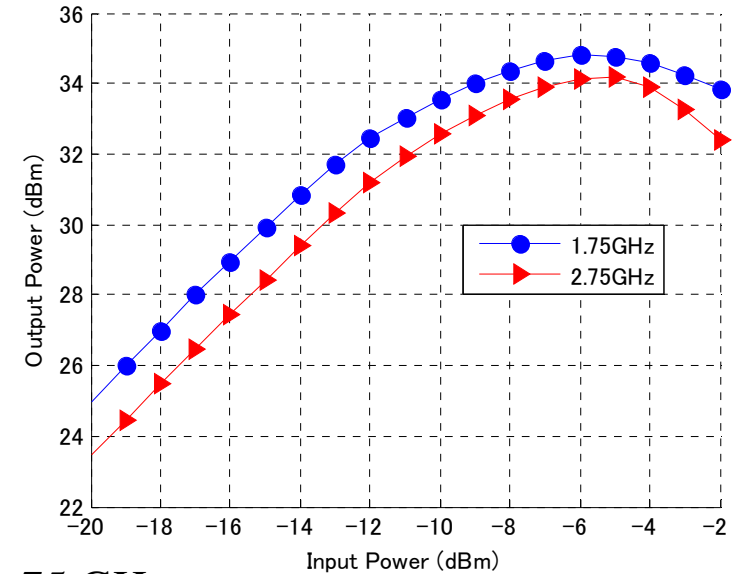




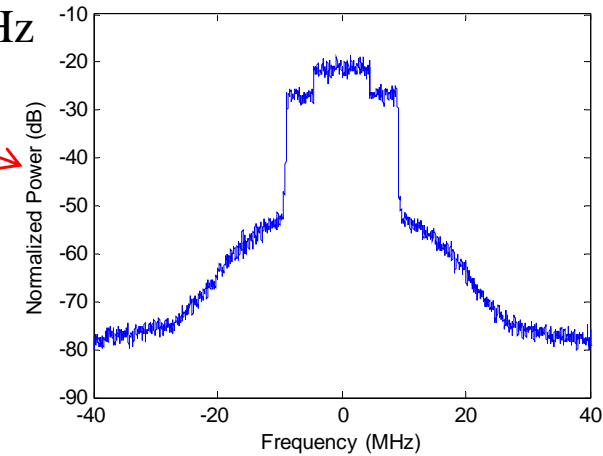
# SFFB Dual-Band DPD by Experiment



### AM-AM characteristics of PA

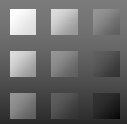


$f_1 = 1.75 \text{ GHz}$   
 $f_2 = 2.75 \text{ GHz}$



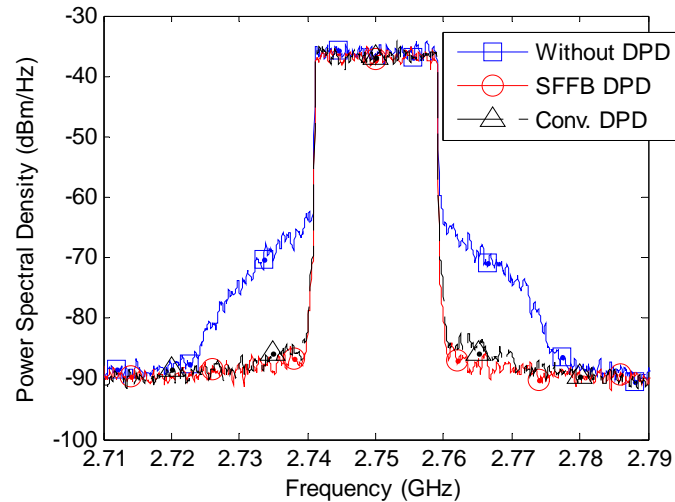
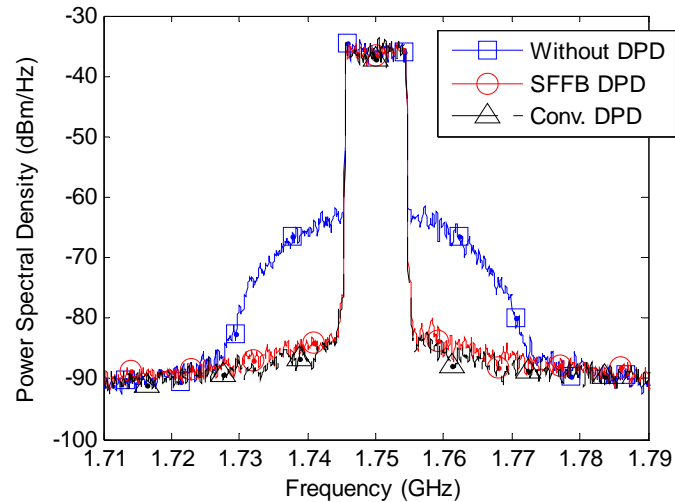
Folded feedback signal without DPD



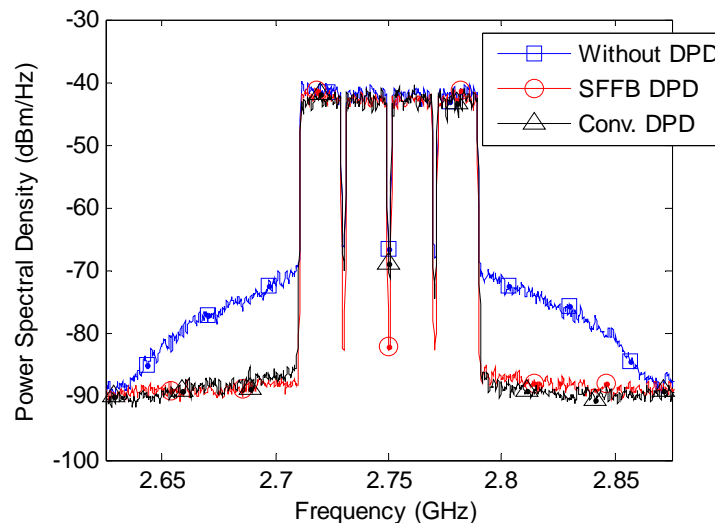
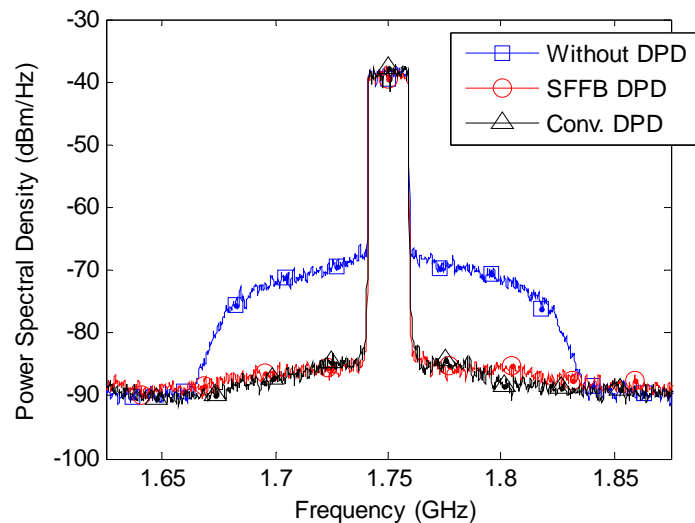


# SFFB Dual-Band DPD by Experiment

Scenario 1; 10 MHz LTE(1.75 GHz) + 20 MHz LTE (2.75 GHz)

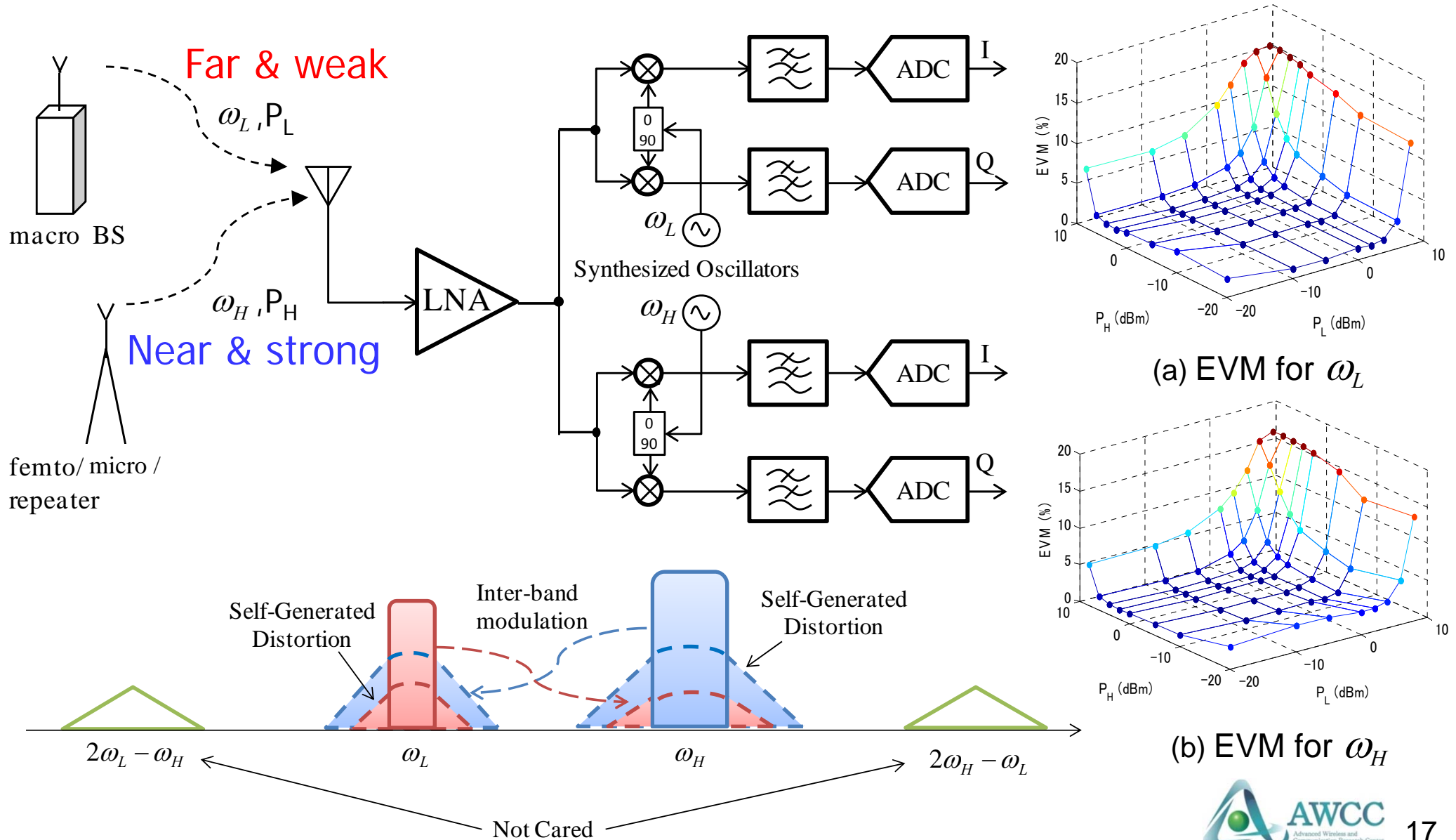


Scenario 2; 20 MHz LTE (1.75 GHz) + 4 x 20 MHz LTE CA (2.75 GHz)

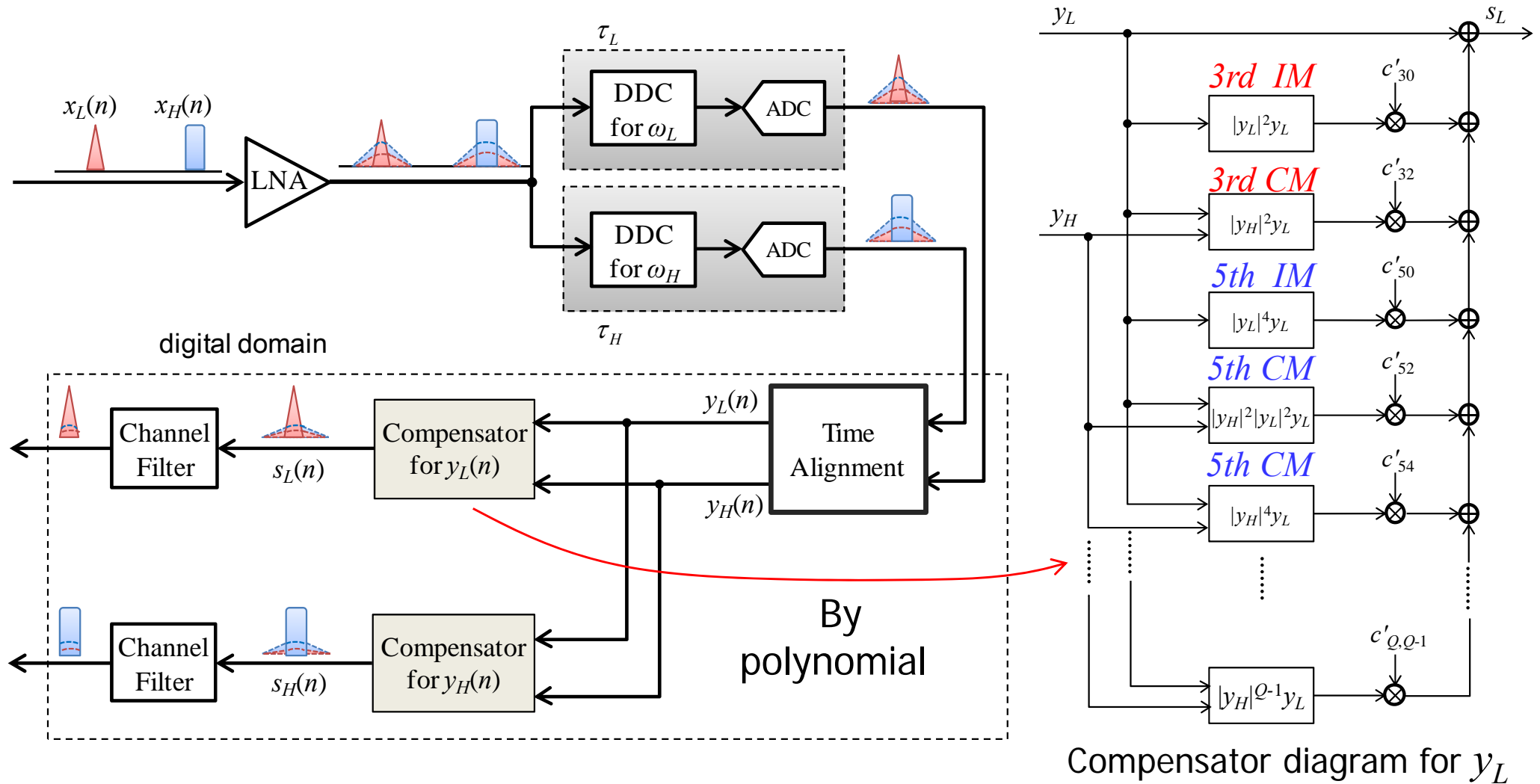


# 3. Cross-Modulation due to Multi-Band Access

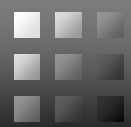
● Near-far problem increases chance of inter-band modulation



# Post-Compensation of Receiver Nonlinearity

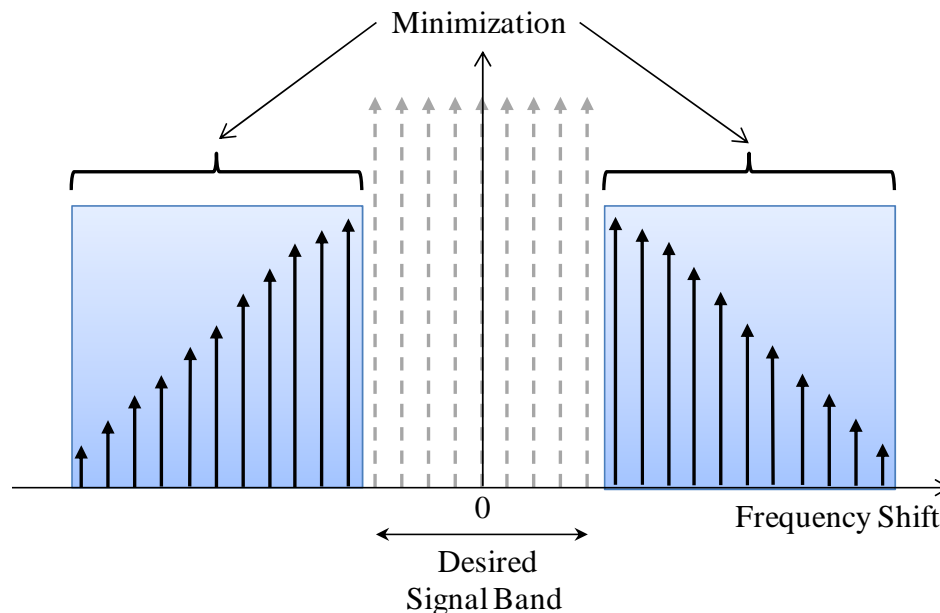


Blind & Adaptive Nonlinear Compensation method is necessary [4].



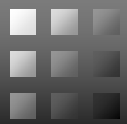
# How to Determine Compensator Coefficients?

- Proposed algorithm determines the compensator coefficients so as to minimize outband spectra power for each band.

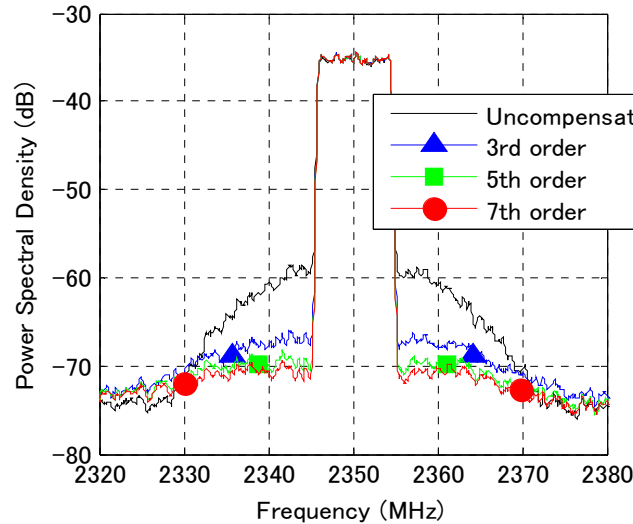


This works successfully because nonlinearity generates both inband and outband components and they are correlated.

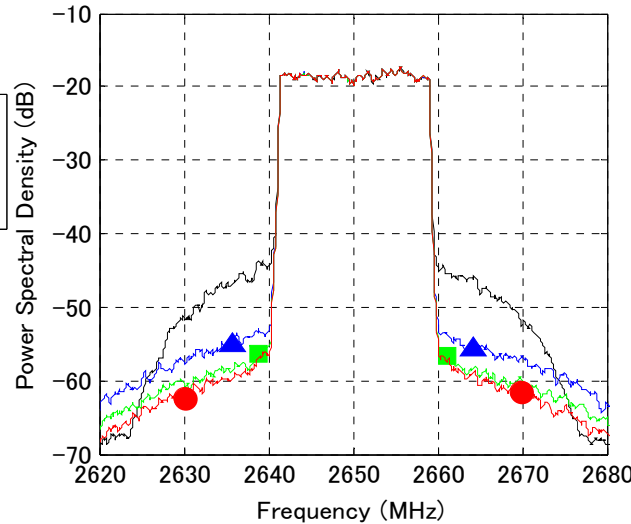
The key is separation filter that eliminates the in-band signal and maximizes SNR of outband distortion detection.



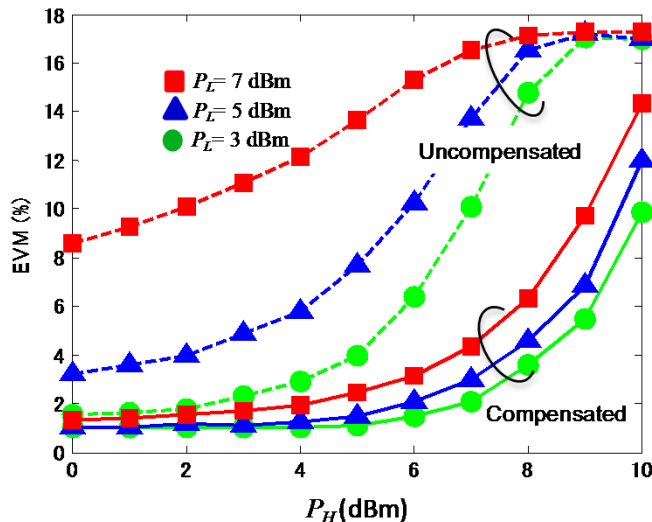
# Experimental Results



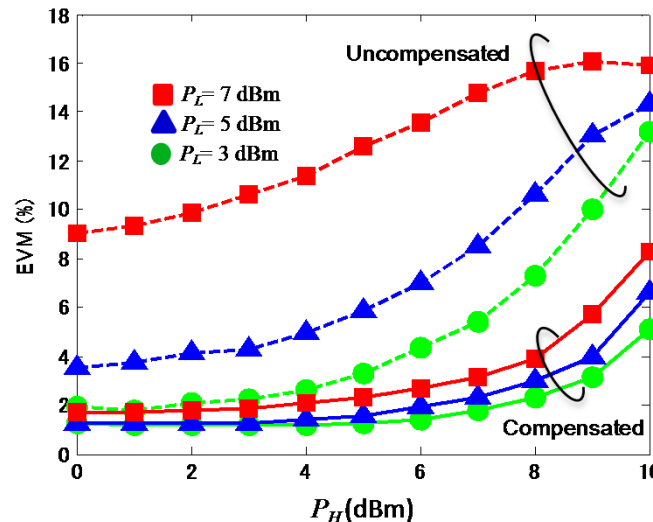
(a) Output spectra at  $\omega_L$



(b) Output spectra at  $\omega_H$



(c) EVM for  $\omega_L$  signal



(d) EVM for  $\omega_H$  signal

$$f_1 = 2.35 \text{ GHz}$$

$$f_2 = 2.65 \text{ GHz}$$

Power levels of  $\omega_L$  and  $\omega_H$  in (a)/(b) are -10 dBm and 6 dBm, respectively.

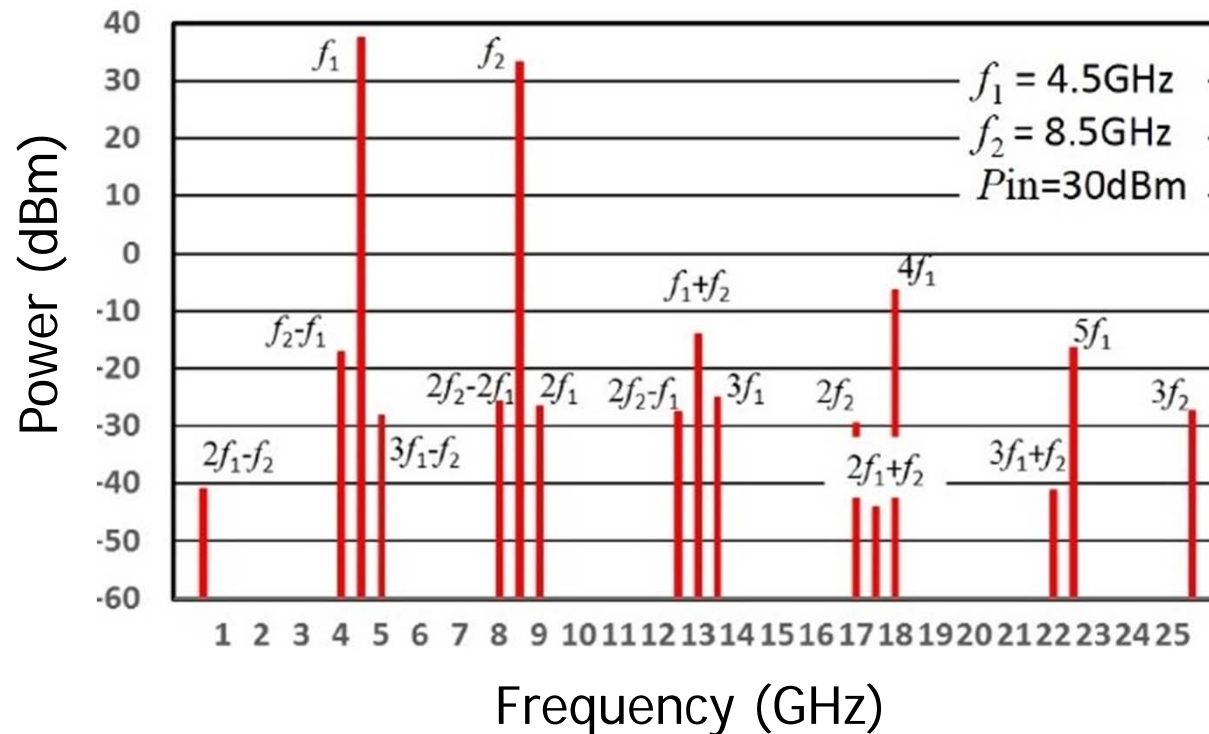
7th-order compensation presents the best results.

Proposed compensator improves EVM in both bands.

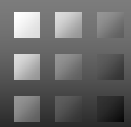
Maximum acceptable input power increases more than 4 dB.

## 4. Reconfigurable RF BPF

- Reconfigurable BPF can adapt the receiver flexibly to the change of access channels and prevent unexpected cross-modulation distortions in multi-band/multi-access operation.

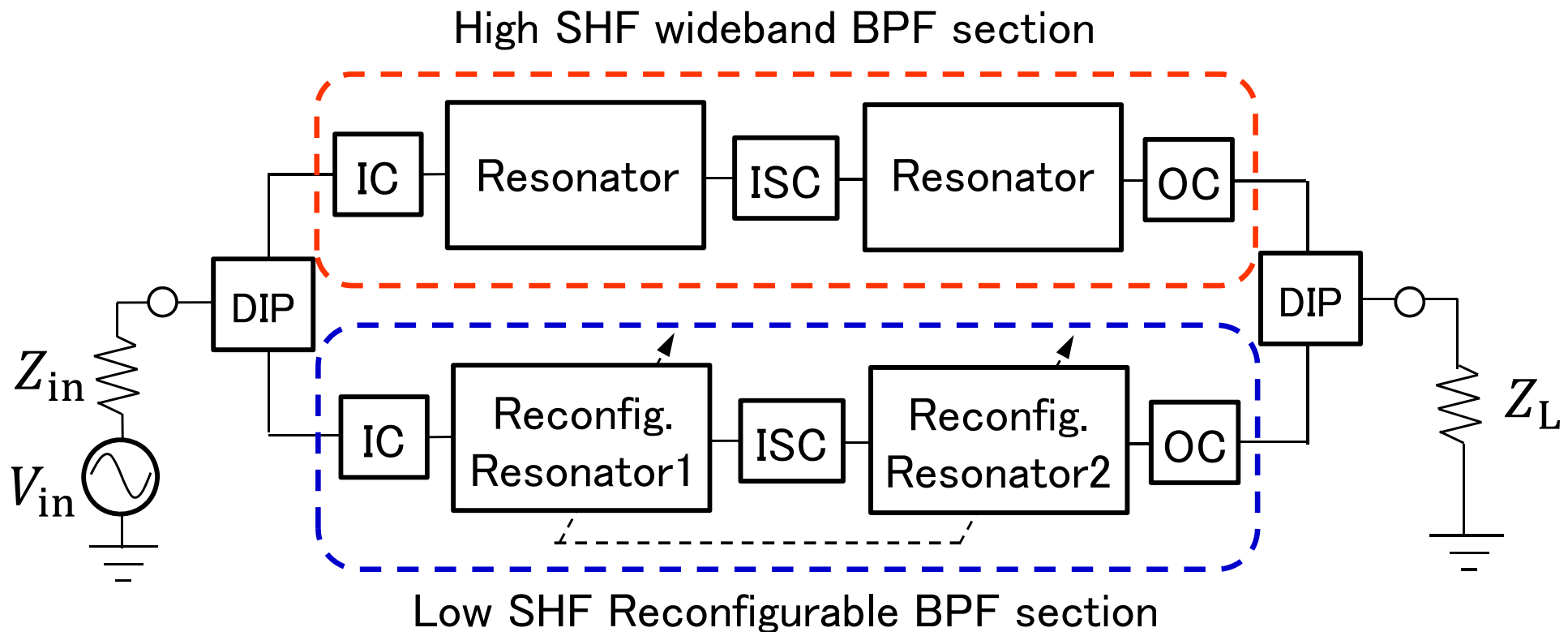


Example of nonlinear output spectrum under concurrent dual-band operation. Some of them locate near the operating band.



# SHF Dual-band Reconfigurable BPF

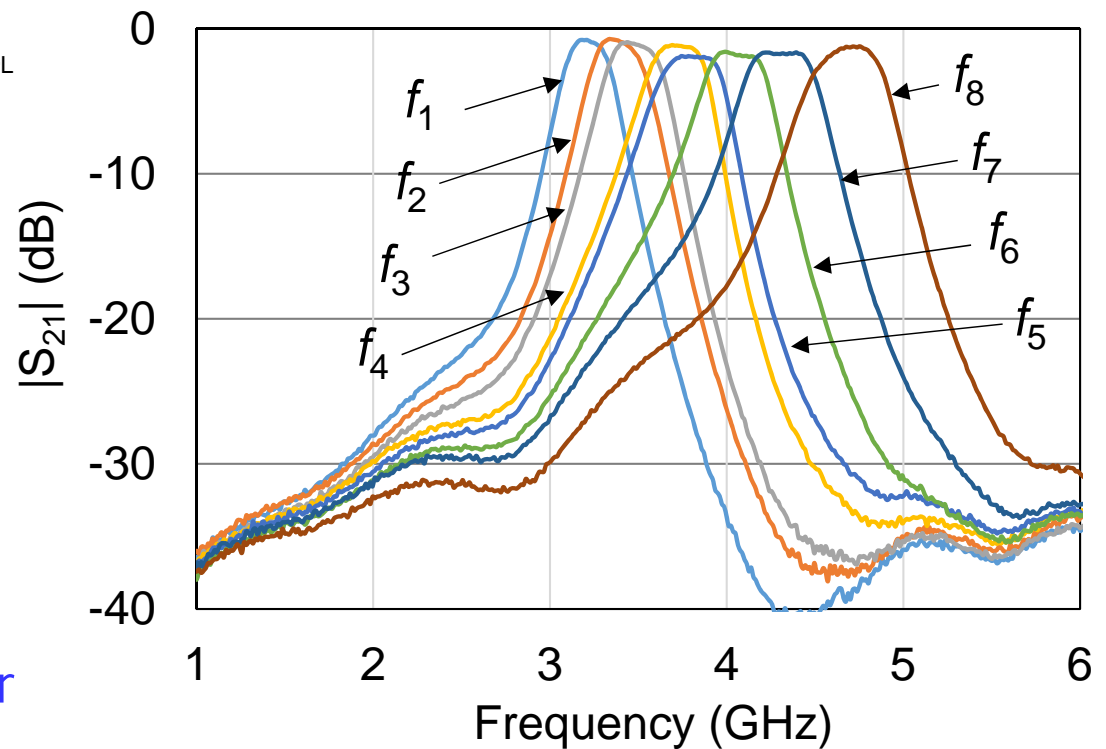
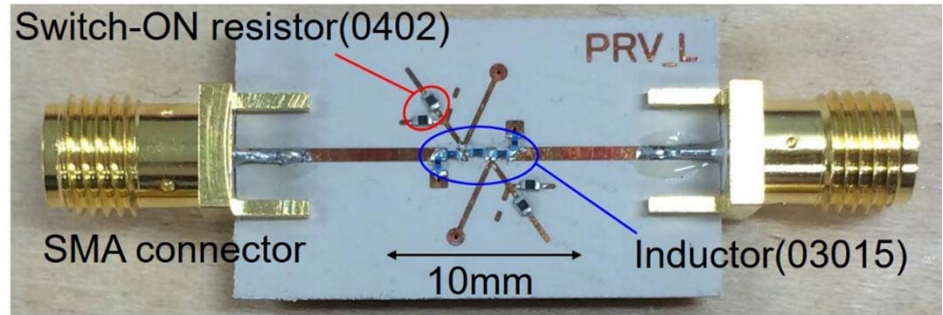
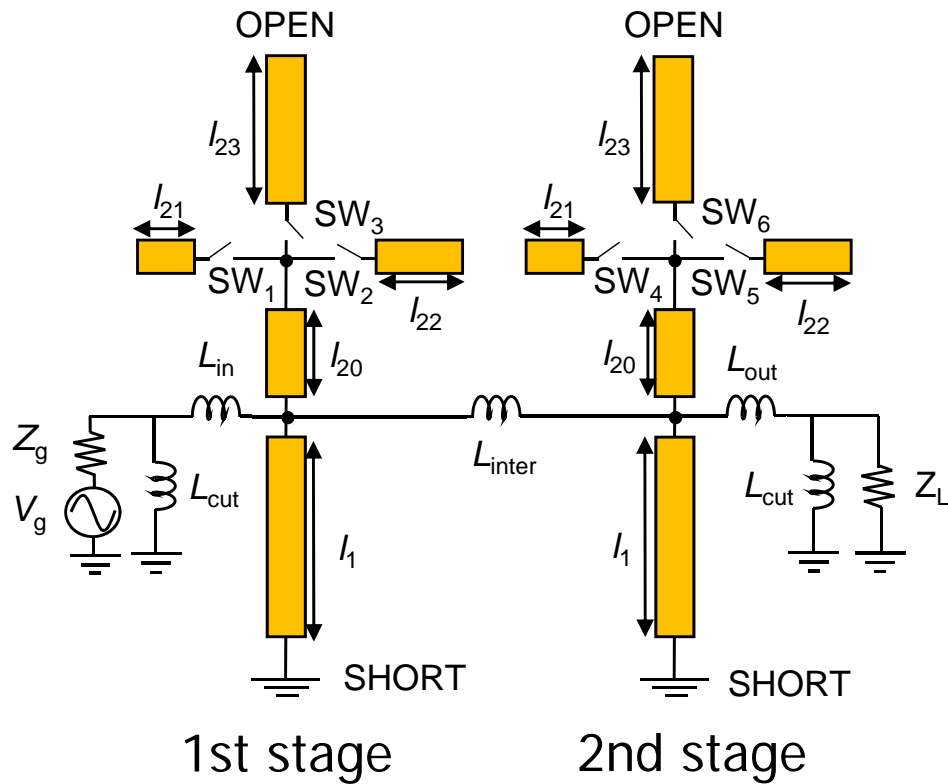
- Low SHF (< 6 GHz) reconfigurable BPF and High SHF wideband BPF are integrated for concurrent dual-band access.



DIP: Diplexer    IC: Input Coupling circuit  
ISC; Inter-Stage Coupling circuit    OC: Output Coupling circuit



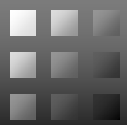
# 2-stage 3-bit Low-SHF Reconfigurable BPF



Center Frequency; 3.1 to 4.6 GHz  
Insertion loss; 1.0~2.1 dB

High SHF BPF and duplexers are under design process.





# Conclusions

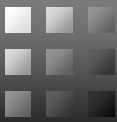
- 5G hardware is required to be more flexible, accurate, linear and low-cost.
- For TxS, advanced DPD techniques called **SENF** and **SFFB** are proposed.
  - SENF saves feedback bandwidth, which enables **wider compensation bandwidth**.
  - SFFB enables simple feedback for **concurrent dual-band operation**.
- For concurrent dual-band RxS, **Post Compensation technique** and **Reconfigurable BPF** are proposed.

The results will contribute to improve practical performance of the 5G system.



# Related Works

- 1) Y. Ma, Y. Yamao, Y. Akaiwa, K. Ishibashi, "Wideband digital predistortion using spectral extrapolation of band-limited feedback signal," IEEE Trans. Circuit and Systems-I, vol. 61, no. 7, pp. 2088-2097, July 2014.
- 2) Y. Ma and Y. Yamao, "Experimental results of digital predistorter for very wideband mobile communication system," Proc. IEEE VTC2015-Spring, 6PB-2, Glasgow, UK, May 2015.
- 3) Y. Ma and Y. Yamao, "Spectra-folding feedback architecture for concurrent dual-band power amplifier predistortion," IEEE Trans. Microw. Theory & Tech., Vol. 63, No. 10, pp. 3164-3174, Oct. 2015.
- 4) Y. Ma, Y. Yamao, K. Ishibashi and Y. Akaiwa, "Adaptive compensation of inter-band modulation distortion for tunable concurrent dual-band receivers," IEEE Trans. Microw. Theory & Tech., vol.61, no.12, pp.4209-4219, Dec. 2013.
- 5) R. Kobayashi, T. Kato, K. Azuma and Y. Yamao, "Design and Fabrication of Two-Stage Three-Bit Reconfigurable Bandpass Filter Using Branch Line-Type Variable Resonator," IEICE Trans. Electronics, Vol. E98-C, No. 7, pp. 636-643, July 2015.



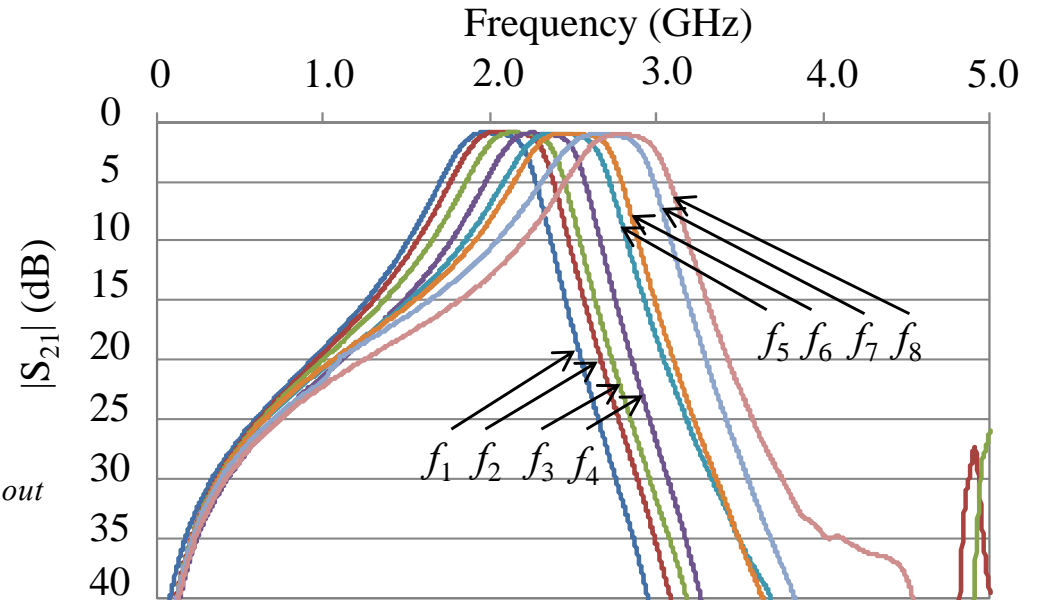
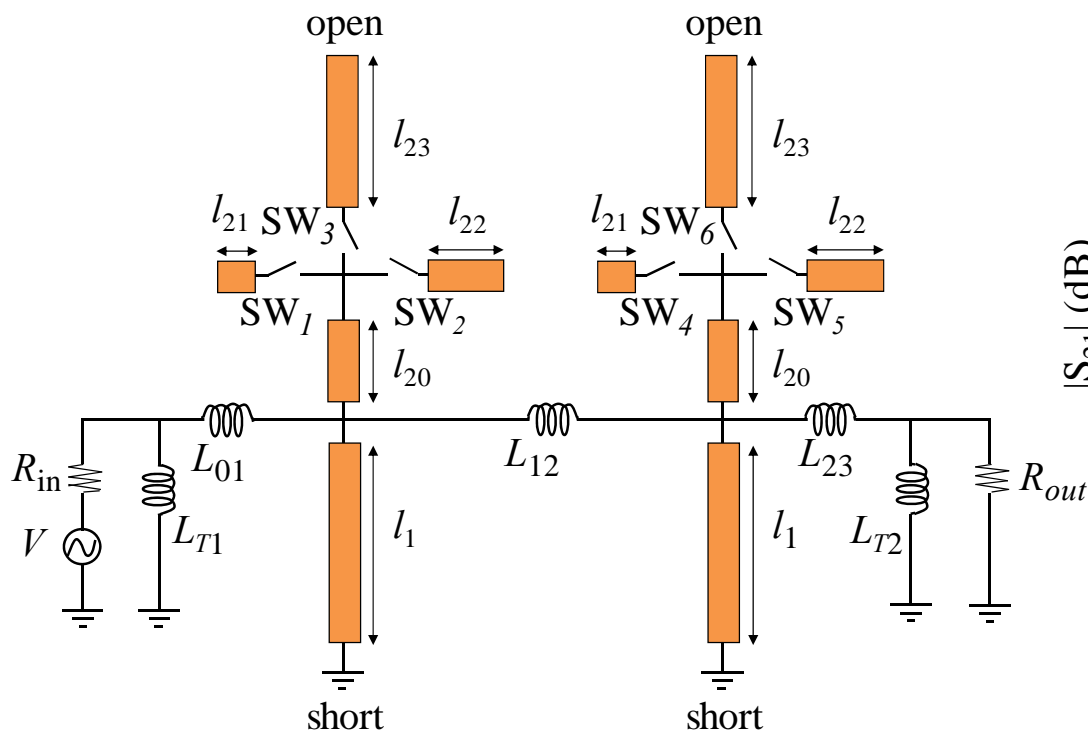
## Acknowledgement

A part of this work is supported by the Ministry of Internal Affairs and Communications (MIC) of Japan under the program "R&D for the realization of 5G mobile communication system".

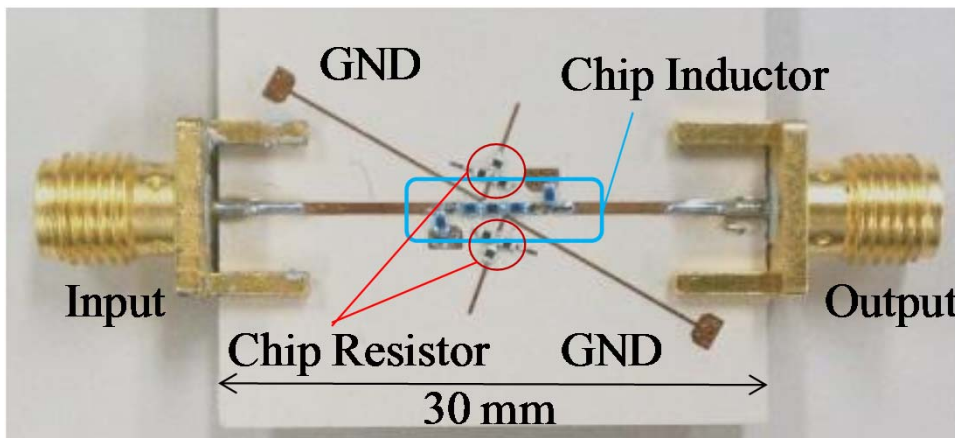
*Thank you for listening!*



# 2-Stage 3-bit Reconfigurable BPF in UHF Band



| $f_i$ | Center frequency (GHz) | Insertion loss (dB) | Band width (MHz) |
|-------|------------------------|---------------------|------------------|
| $f_1$ | 2.01                   | 0.90                | 505              |
| $f_2$ | 2.09                   | 0.81                | 542              |
| $f_3$ | 2.13                   | 0.90                | 520              |
| $f_4$ | 2.24                   | 0.92                | 528              |
| $f_5$ | 2.38                   | 0.94                | 548              |
| $f_6$ | 2.44                   | 0.93                | 552              |
| $f_7$ | 2.72                   | 0.96                | 574              |
| $f_8$ | 2.80                   | 0.98                | 567              |



(From APMC2012, Kaohsiung Taiwan)